

## DIO6012

# High-Efficiency 1 MHz, 2 A Output Synchronous Step Down Converter

## Features

- Low  $R_{DS(ON)}$  for internal switches (top/bottom)  
100 mΩ / 80 mΩ, 2.0 A
- 2.7-5.5 V input voltage range
- High switching frequency minimizes the external components
- Internal soft-start limits the inrush current
- 100% dropout operation capability with input voltage up to 5 V
- Green package:  
SOT23-6, SOT23-5 is pin-compatible.
- Other packages are available upon requests
- No latch off output voltage protection

## Applications

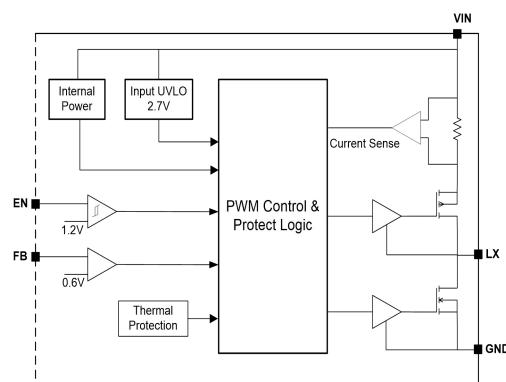
- LCD TV
- Set top box
- Net PC
- Mini-notebook PC
- Access point router

## Descriptions

The DIO6012 is high-efficiency, high frequency synchronous step-down DC-DC regulator ICs capable of delivering up to 2 A output currents. The DIO6012 family operate over a wide input voltage range from 2.7 V to 5.5 V and integrate main switch and synchronous switch with very low  $R_{DS(ON)}$  to minimize the conduction loss.

Low output voltage ripple and small external inductor and capacitor sizes are achieved with greater than 1 MHz switching frequency.

## Function Block



## Ordering Information

Order Part No.	Top Marking	RoHS	MSL	$T_A$	Package	
DIO6012CST6	CYWC	Green	MSL3	-40 to 85°C	SOT23-6	Tape & Reel, 3000
DIO6012CST5	CYWB	Green	MSL3	-40 to 85°C	SOT23-5	Tape & Reel, 3000

### Marking Definition



## Pin Assignments

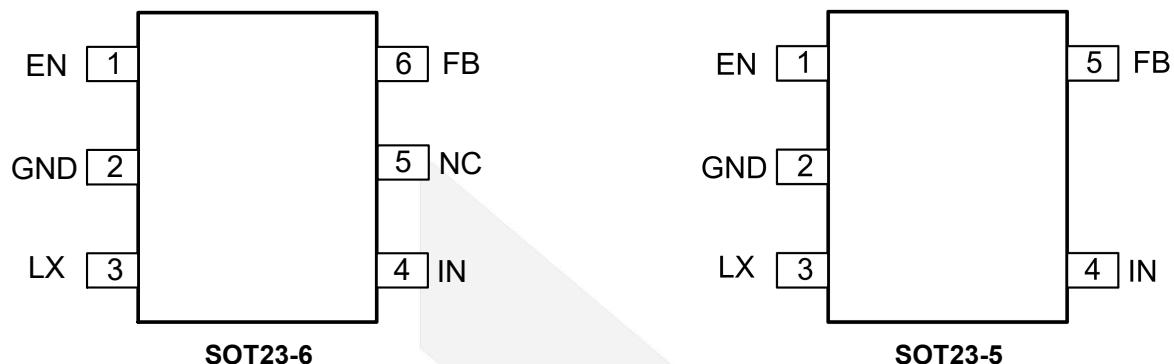


Figure 1. Pin assignment (Top view)

## Pin Definitions

Pin Name	Description
EN	Enable control. Pull high to turn on. Do not float.
GND	Ground
LX	Inductor pin. Connect this pin to the switching node of inductor.
IN	Power input
NC	No connect
FB	Output feedback pin. Connect this pin to the center point of the output resistor divider (as shown in Figure 1) to program the output voltage: $V_{OUT} = 0.6*(1+R1/R2)$ .

**DIO6012**

## Absolute Maximum Ratings

Stresses beyond those listed under "Absolute Maximum Rating" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other condition beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Symbol	Parameter	Rating	Unit
$V_{IN}$	Supply voltage ( $V_+ - V_-$ )	-0.3 to 6.0	V
	Enable, FB voltage	-0.3 to $V_{IN}+0.6$	V
$P_D$	Power dissipation, $P_D$ at $T_A = 25^\circ\text{C}$ , SOT23-6	0.6	W
$\theta_{JA}$	Package thermal resistance	170	$^\circ\text{C}/\text{W}$
$\theta_{JC}$		130	
$T_{STG}$	Storage temperature range	-65 to 150	$^\circ\text{C}$
$T_J$	Junction temperature range	150	$^\circ\text{C}$
$T_L$	Lead temperature range	260	$^\circ\text{C}$
ESD	HBM, JEDEC: JESD22-A114	6000	V
	MM, JEDEC: JESD22-A115	200	
	Dynamic LX voltage in 50 ns duration	$V_{IN}+3$ to GND-4	V

## Recommend Operating Conditions

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended Operating conditions are specified to ensure optimal performance to the datasheet specifications. DIOO does not recommend exceeding them or designing to Absolute Maximum Ratings.

Symbol	Parameter	Rating	Unit
$V_{IN}$	Supply voltage	2.7 to 5.5	V
$T_J$	Junction temperature range	-40 to 125	$^\circ\text{C}$
$T_A$	Ambient temperature range	-40 to 85	$^\circ\text{C}$



DIO6012

## Electrical Characteristics

$V_{IN} = 3.6 \text{ V}$ ,  $V_{OUT} = 1.8 \text{ V}$ ,  $L = 2.2 \mu\text{H}$ ,  $C_{OUT} = 22 \mu\text{F}$ ,  $T_A = 25^\circ\text{C}$ , unless otherwise specified.

Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
$V_{IN}$	Input voltage range		2.7		5.5	V
$I_Q$	Quiescent current	$I_{OUT} = 0$ , $V_{FB} = V_{REF} \cdot 105\%$		80		$\mu\text{A}$
$I_{SHDN}$	Shutdown current	$EN = 0$		0.1	1	$\mu\text{A}$
$V_{REF}$	Feedback reference voltage		0.588	0.6	0.612	V
$I_{FB}^{(1)}$	FB input current	$V_{FB} = V_{IN}$	-50		50	nA
$R_{DS(ON),P}^{(1)}$	PFET $R_{ON}$			100		$\text{m}\Omega$
$R_{DS(ON),N}^{(1)}$	NFET $R_{ON}$			80		$\text{m}\Omega$
$I_{LIM}^{(1)}$	PFET current limit		2.5			A
$V_{ENH}$	EN rising threshold		1.2			V
$V_{ENL}$	EN falling threshold				0.4	V
$V_{UVLO}$	Input UVLO threshold				2.7	V
$V_{HYS}$	UVLO hysteresis			0.1		V
$V_{OVP}$	Over voltage protection threshold		0.69	0.72	0.75	V
$F_{osc}$	Oscillator frequency	$I_{OUT} = 500 \text{ mA}$		1		MHz
	Min ON time <sup>(1)</sup>			50		ns
	Max duty cycle		100			%
$T_{SD}^{(1)}$	Thermal shutdown temperature			150		$^\circ\text{C}$

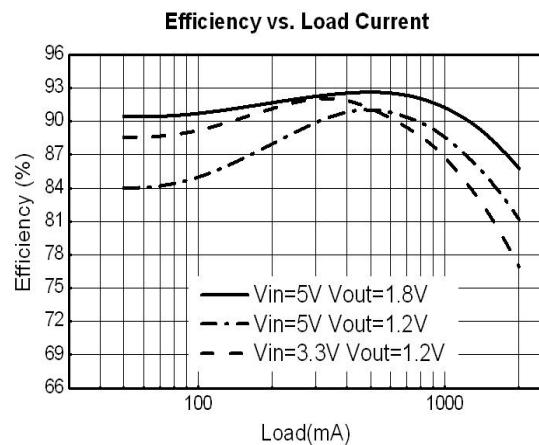
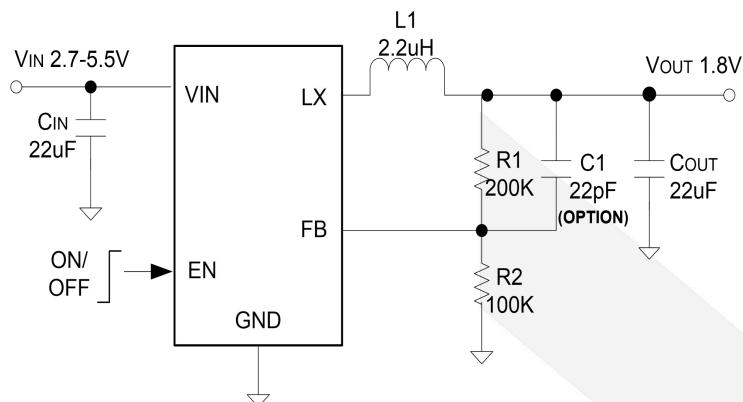
**Note:**

(1) Guaranteed by design.

(2) Specifications subject to change without notice.

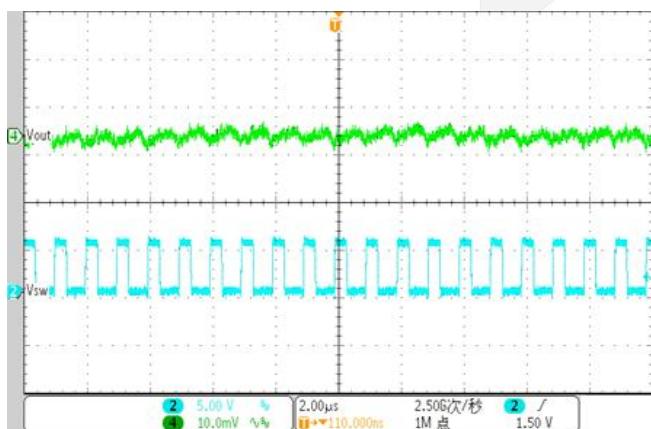
High-Efficiency 1 MHz, 2 A Output Synchronous Step Down Converter

## Typical Application

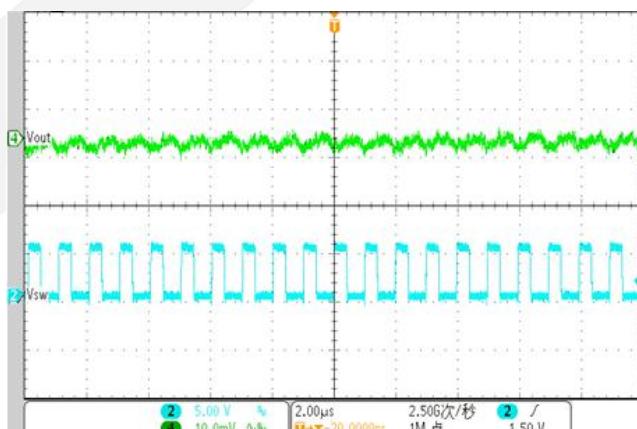


## Typical Performance Characteristic

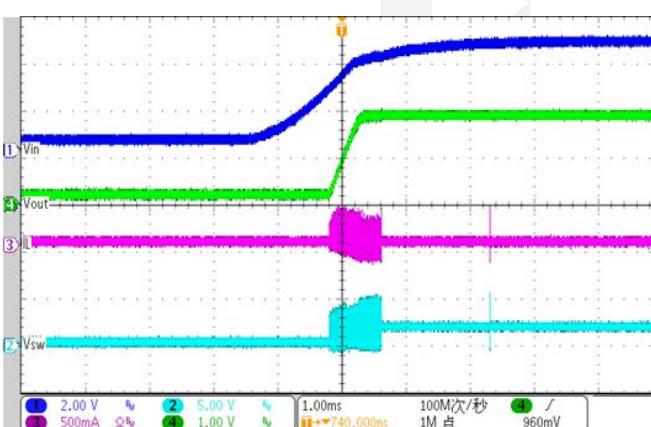
$V_{IN} = 5$  V,  $V_{OUT} = 1.8$  V,  $L = 2.2 \mu$ H,  $C_{OUT} = 22 \mu$ F,  $T_A = 25^\circ$ C, unless otherwise specified.



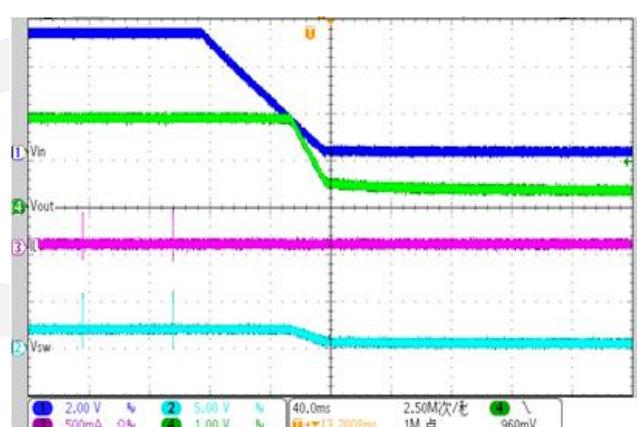
**Ripple**  
( $V_{IN} = 5$  V,  $V_{OUT} = 1.8$  V, Load = 1 A)



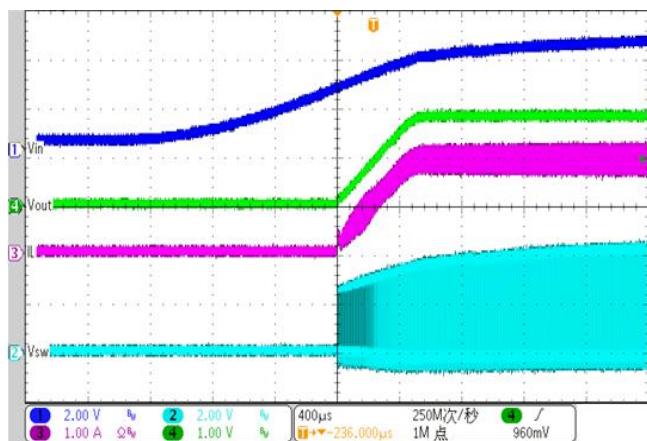
**Ripple**  
( $V_{IN} = 5$  V,  $V_{OUT} = 1.8$  V, Load = 2 A)



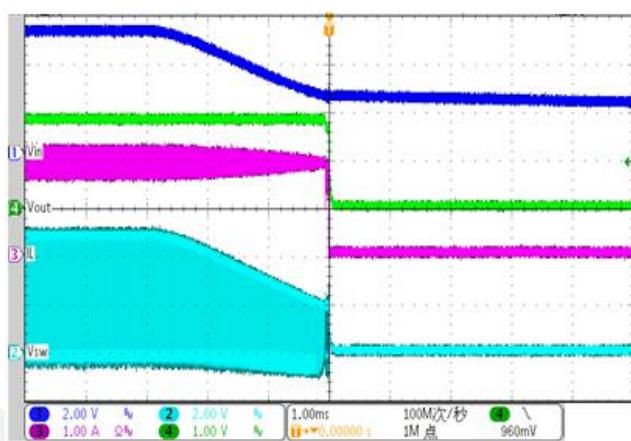
**$V_{IN}$  start up**  
( $V_{IN} = 5$  V,  $V_{OUT} = 1.8$  V, Load = 0 A)



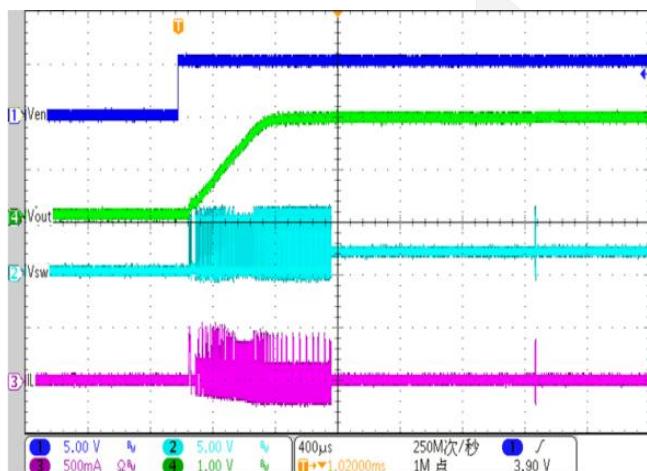
**$V_{IN}$  shut down**  
( $V_{IN} = 5$  V,  $V_{OUT} = 1.8$  V, Load = 0 A)



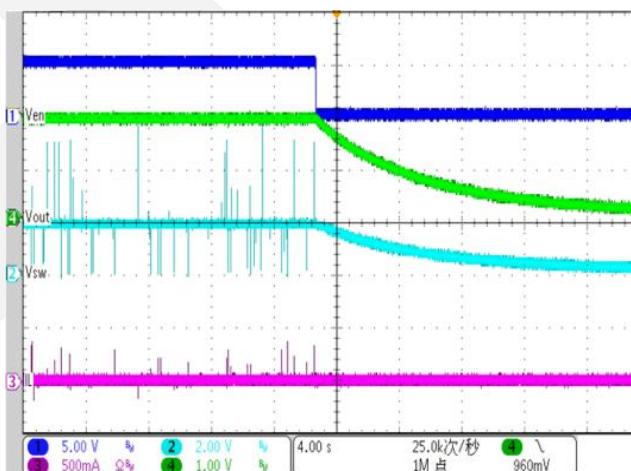
**V<sub>IN</sub> start up**  
(V<sub>IN</sub> = 5 V, V<sub>OUT</sub> = 1.8 V, Load = 2 A)



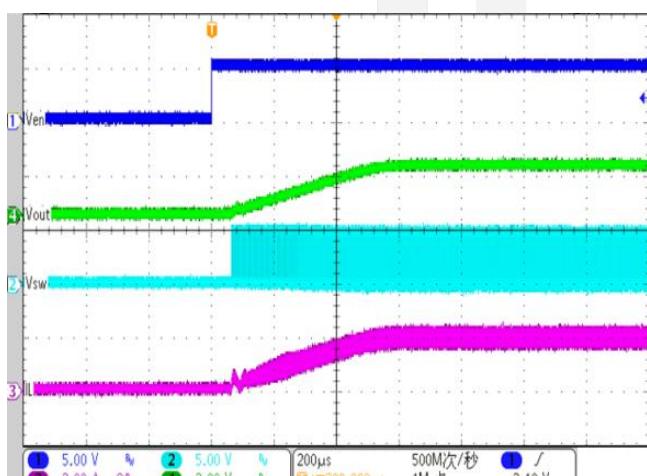
**V<sub>IN</sub> shut down**  
(V<sub>IN</sub> = 5 V, V<sub>OUT</sub> = 1.8 V, Load = 2 A)



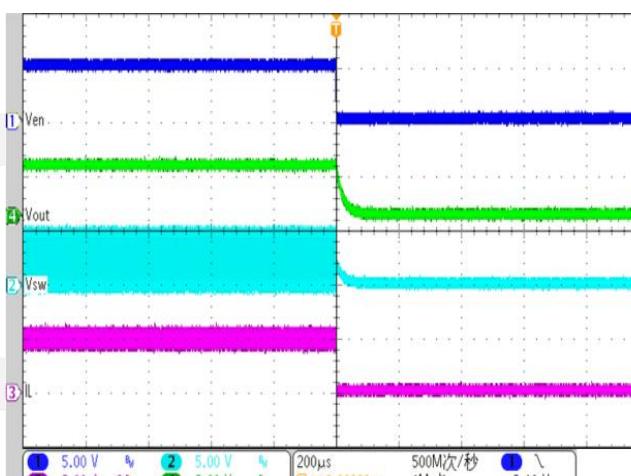
**Enable start up**  
(V<sub>IN</sub> = 5 V, V<sub>OUT</sub> = 1.8 V, Load = 0 A)



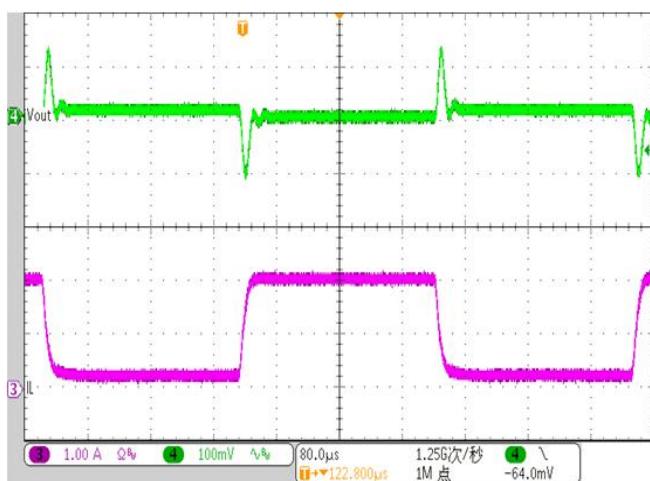
**Enable shut down**  
(V<sub>IN</sub> = 5 V, V<sub>OUT</sub> = 1.8 V, Load = 0 A)



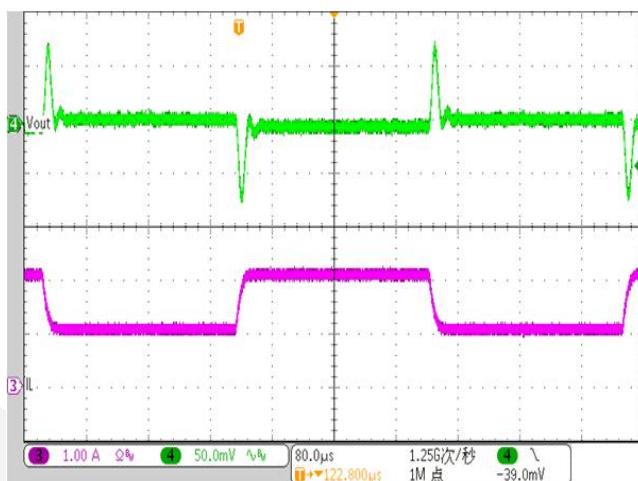
**Enable start up**  
(V<sub>IN</sub> = 5 V, V<sub>OUT</sub> = 1.8 V, Load = 2 A)



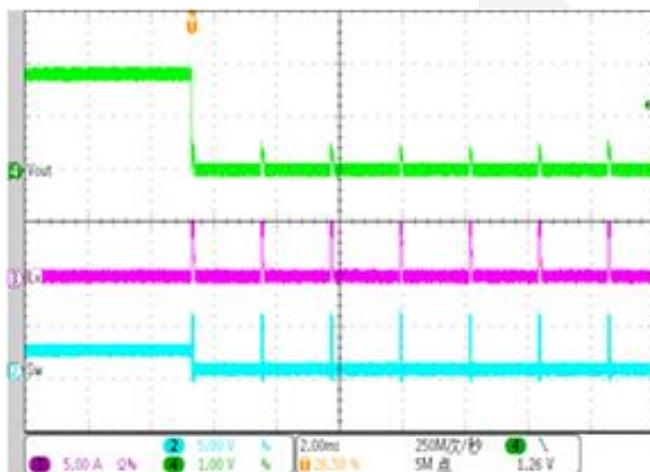
**Enable shut down**  
(V<sub>IN</sub> = 5 V, V<sub>OUT</sub> = 1.8 V, Load = 2 A)



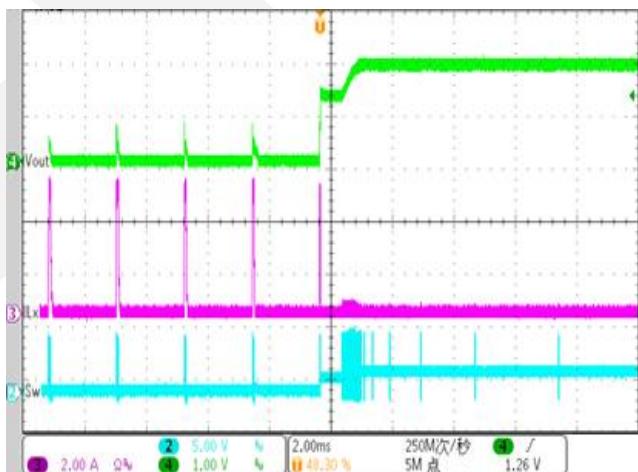
**Transient**  
 $(V_{IN} = 5 \text{ V}, V_{OUT} = 1.8 \text{ V}, \text{Load} = 0.2-2 \text{ A}, 1.6 \text{ A/us})$



**Transient**  
 $(V_{IN} = 5 \text{ V}, V_{OUT} = 1.8 \text{ V}, \text{Load} = 1-2 \text{ A}, 1.6 \text{ A/us})$



**Short protection**  
 $(V_{IN} = 5 \text{ V}, V_{OUT} = 1.8 \text{ V}, \text{No Load} \rightarrow \text{short})$



**Short recovery**  
 $(V_{IN} = 5 \text{ V}, V_{OUT} = 1.8 \text{ V}, \text{short} \rightarrow \text{No Load})$

## Application Information

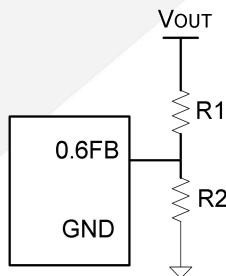
DIO6012 is a synchronous buck regulator IC that integrates the PWM control, top and bottom switches on the same die to minimize the switching transition loss and conduction loss. With ultra low  $R_{DS(ON)}$  power switches and proprietary PWM control, this regulator IC can achieve the highest efficiency and the highest switch frequency simultaneously to minimize the external inductor and capacitor size, and thus achieving the minimum solution footprint.

Because of the high integration in the DIO6012 IC, the application circuit based on this regulator IC is rather simple. Only input capacitor  $C_{IN}$ , output capacitor  $C_{OUT}$ , output inductor L and feedback resistors (R1 and R2) need to be selected for the targeted applications specifications.

### Feedback resistor dividers R1 and R2

Choose R1 and R2 to program the proper output voltage. To minimize the power consumption under light loads, it is desirable to choose large resistance values for both R1 and R2. A value of between 10k and 1M is highly recommended for both resistors. If  $V_{OUT}$  is 1.8 V, R1 = 100 k is chosen, then R2 can be calculated to be 50k.

$$R_2 = \frac{0.6V}{V_{OUT} - 0.6V} R_1 \quad (1)$$



### Input capacitor $C_{IN}$

This ripple current through input capacitor is calculated as:

$$I_{CIN\_RMS} = I_{OUT} \cdot \sqrt{D(1 - D)} \quad (2)$$

This formula has a maximum at  $V_{IN} = 2 V_{OUT}$  condition, where  $I_{CIN\_RMS} = I_{OUT}/2$ . This simple worst-case condition is commonly used for DC/DC design.

With the maximum load current at 1.2 A. A typical X5R or better grade ceramic capacitor with 6 V rating and more than 1pcs 10 uF capacitor can handle this ripple current well. To minimize the potential noise problem, place this ceramic capacitor really close to the IN and GND pins. Care should be taken to minimize the loop area formed by  $C_{IN}$ , and IN/GND pins.



# DIO6012

## High-Efficiency 1 MHz, 2 A Output Synchronous Step Down Converter

### Output capacitor $C_{OUT}$

The output capacitor is selected to handle the output ripple noise requirements. Both steady state ripple and transient requirements must be taken into consideration when selecting this capacitor. For the best performance, it is recommended to use X5R or better grade ceramic capacitor with 6 V rating and greater than 40  $\mu$ F capacitance.

### Output inductor $L$ :

There are several considerations in choosing this inductor.

- 1) Choose the inductance to provide the desired ripple current. It is suggested to choose the ripple current to be about 40% of the maximum output current. The inductance is calculated as:

$$L = \frac{V_{OUT}(1 - V_{OUT}/V_{IN, MAX})}{F_{SW} \times I_{OUT, MAX} \times 40\%} \quad (3)$$

where  $F_{SW}$  is the switching frequency and  $I_{OUT, MAX}$  is the maximum load current. The DIO6012 regulator IC is quite tolerant of different ripple current amplitude. Consequently, the final choice of inductance can be slightly off the calculation value without significantly impacting the performance.

- 2) The saturation current rating of the inductor must be selected to be greater than the peak inductor current under full load conditions.

$$I_{SAT, MIN} > I_{OUT, MAX} + \frac{V_{OUT}(1 - V_{OUT}/V_{IN, MAX})}{2 \cdot F_{SW} \cdot L} \quad (4)$$

- 3) The DCR of the inductor and the core loss at the switching frequency must be low enough to achieve the desired efficiency requirement. It is desirable to choose an inductor with  $DCR < 50 \text{ m}\Omega$  to achieve a good overall efficiency.

### Enable Operation

Pulling the EN pin low ( $<0.4\text{V}$ ) will shut down the device. During shut down mode, the DIO6012 shutdown current drops to lower than  $0.1 \mu\text{A}$ . Driving the EN pin high ( $>1.2 \text{ V}$ ) will turn on the IC again.

**Load Transient Considerations:** The DIO6012 regulator IC integrates the compensation components to achieve good stability and fast transient responses. In some applications, adding a  $22 \text{ pF}$  ceramic cap in parallel with R1 may further speed up the load transient responses and is thus recommended for applications with large load transient step requirements.

### Layout Design:

The layout design of DIO6012 regulator is relatively simple. For the best efficiency and minimum noise problems, we should place the following components close to the IC:  $C_{IN}$ ,  $L$ ,  $R1$  and  $R2$ .

- 1) It is desirable to maximize the PCB copper area connecting to GND pin to achieve the best thermal and noise performance. If the board space allowed, a ground plane is highly desirable.
- 2)  $C_{IN}$  must be close to Pins IN and GND. The loop area formed by  $C_{IN}$  and GND must be minimized.
- 3) The PCB copper area associated with LX pin must be minimized to avoid the potential noise problem.

- 4) The components R1 and R2, and the trace connecting to the FB pin must NOT be adjacent to the LX net on the PCB layout to avoid the noise problem.
- 5) If the system chip interfacing with the EN pin has a high impedance state at shutdown mode and the IN pin is connected directly to a power source such as a Li-Ion battery, it is desirable to add a pull down 1 MΩ resistor between the EN and GND pins to prevent the noise from falsely turning on the regulator at shutdown mode.

#### Output voltage ripple test

A proper output ripple measurement should be done according to Figure 5 setup. Output voltage ripple should be measured across the output ceramic cap near the IC.

1. Remove the ground clip and head of the probe. Wind thin wires around the ground ring of the probe. Solder the end of the ground ring wire to the negative node of the  $C_{OUT}$ . Touch the probe tip to the positive node of the  $C_{OUT}$ . (Refer to Figure.5).
2. Minimize the loop formed by  $C_{OUT}$  terminals, probe tip and ground ring.
3. Change the probing direction to decouple the electromagnetic noise generated from the nearby buck inductor. (Refer to Figure.5)

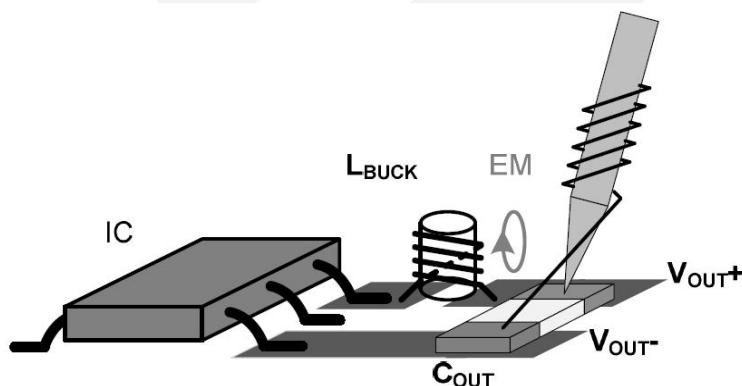
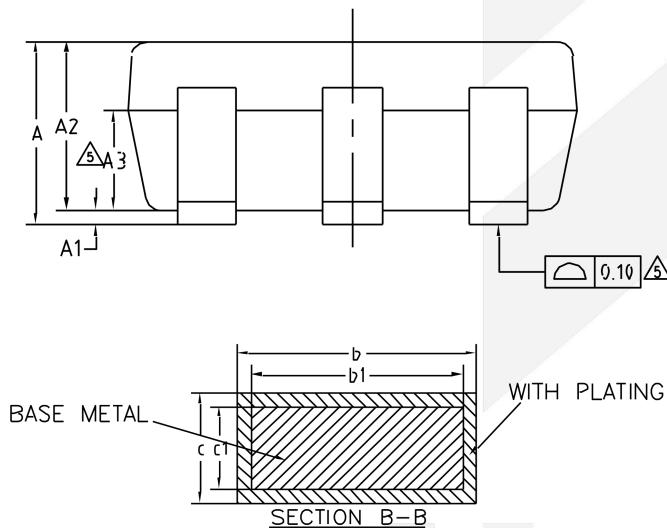
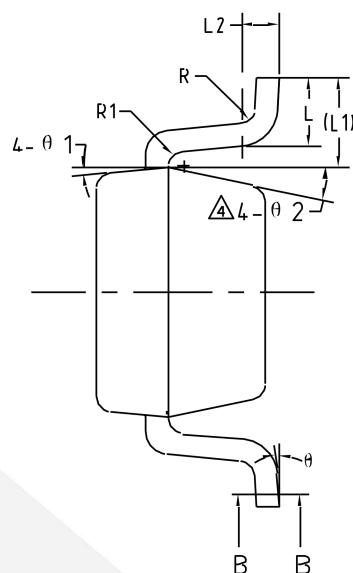
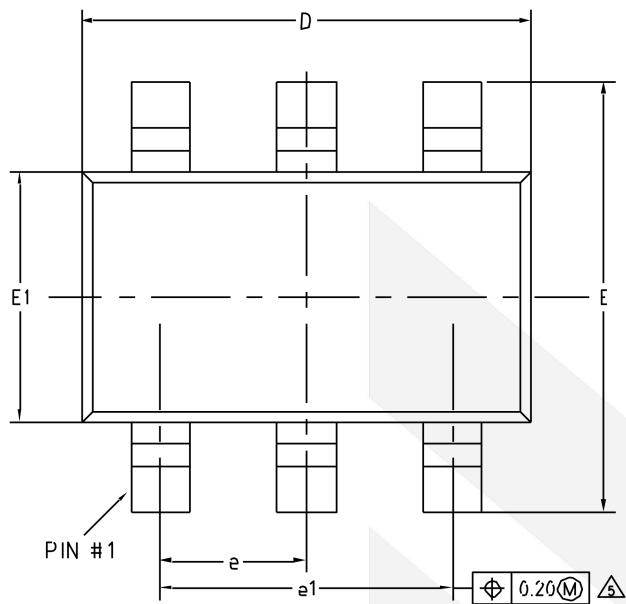


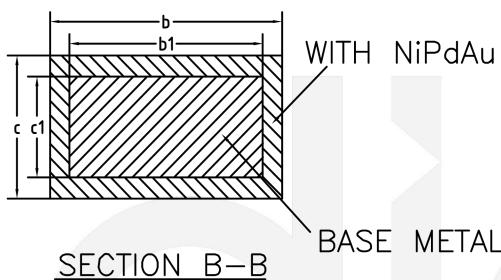
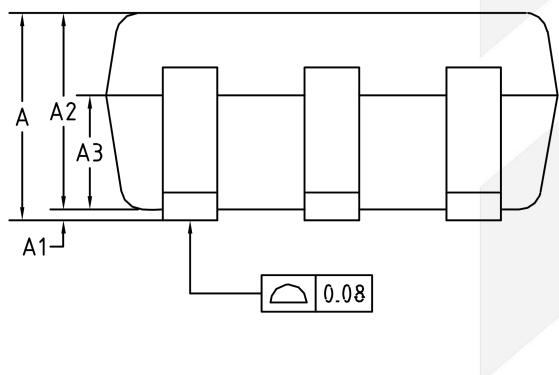
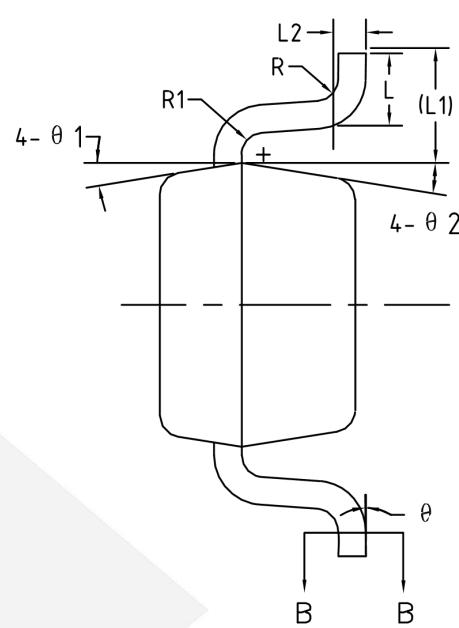
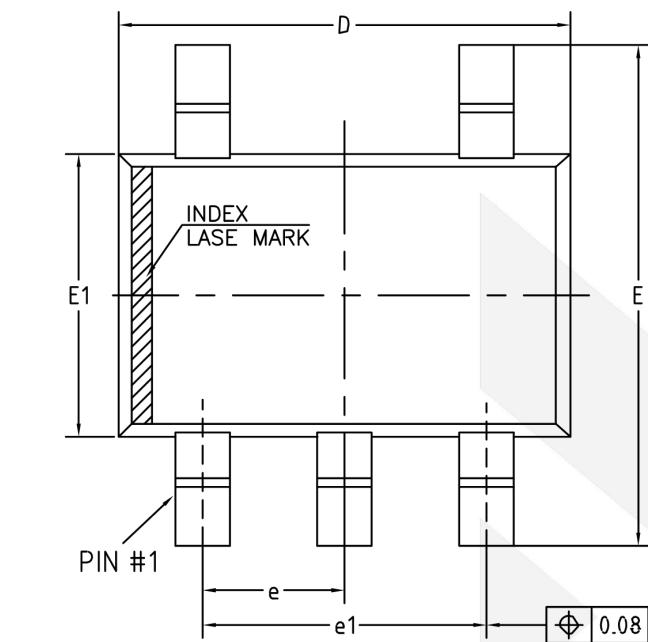
Figure 5. Recommended way to measure the output voltage ripple

## Physical Dimensions: SOT23-6



Common Dimensions (Units of measure = Millimeter)			
Symbol	Min	Nom	Max
A	-	-	1.25
A1	0	-	0.15
A2	1.00	1.10	1.20
A3	0.60	0.65	0.70
b	0.36	-	0.50
b1	0.36	0.38	0.45
c	0.14	-	0.20
c1	0.14	0.15	0.16
D	2.826	2.926	3.026
E	2.60	2.80	3.00
E1	1.526	1.626	1.726
e	0.90	0.95	1.00
e1	1.80	1.90	2.00
L	0.35	0.45	0.60
L1	0.59 REF		
L2	0.25 BSC		
R	0.10	-	-
R1	0.10	-	0.25
θ	0°	-	8°
θ1	3°	5°	7°
θ2	6°	-	14°

## Physical Dimensions: SOT23-5



Common Dimensions (Units of measure = Millimeter)			
Symbol	Min	Nom	Max
A	-	-	1.25
A1	0	-	0.15
A2	1.00	1.10	1.20
A3	0.60	0.65	0.70
b	0.36	-	0.50
b1	0.36	0.38	0.45
c	0.14	-	0.20
c1	0.14	0.15	0.16
D	2.826	2.926	3.026
E	2.60	2.80	3.00
E1	1.526	1.626	1.726
e	0.90	0.95	1.00
e1	1.80	1.90	2.00
L	0.35	0.45	0.60
L1	0.59 REF		
L2	0.25 BSC		
R	0.10	-	-
R1	0.10	-	0.25
Θ	0°	-	8°
Θ1	3°	5°	7°
Θ2	6°	-	14°



## CONTACT US

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